

# Design and Implementation of GABOR Type Filter on FPGA



## Engineering

**KEYWORDS :** gabor filter, floating point, field programmable gate arrays(FPGAs)

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### ABSTRACT

The gabor filter architecture used in this paper is suitable for real time applications with high pixel rates. The structure is capable of processing images using floating point computation. Xilinx ISE suite is used for simulation. Verilog HDL is used for mapping the algorithm in VLSI and FPGA implementation is done on SPARTAN- 3E.

### Introduction

In image processing Gabor Filters are linear filters mainly used for edge detection. The frequency and orientation selectivity property of these filters are similar to the human visual system. In this brief we are using the floating point computation technique for the design of gabor filters. Floating point computation is more advantageous than any other computation technique, because of its accuracy.

In this paper we will discuss about gabor filters, floating point conversion of numbers, multiplication and addition of floating point numbers used in the basic gabor filter architecture. Finally the simulation is done using Xilinx and FPGA implementation using SPARTAN 3E.

### Gabor filters

Gabor filters are used for edge detection because of its frequency and orientation selectivity properties. In spatial domain, a gabor filter is a Gaussian kernel function modulated by a sinusoidal plane wave. The simple cells in the visual cortex system of humans can be modulated using gabor functions. So image analysis of gabor filters are similar to perception of human eyes.

### Definition of Gabor Filter

The impulse response of gabor filter can be represented by complex valued convolution kernels.

$$g(x,y) = \frac{1}{2\pi\sigma^2} e^{-\frac{x^2+y^2}{2\sigma^2}} e^{j(\omega_x x + \omega_y y)}$$

$\sigma$ ,  $\omega_x$  and  $\omega_y$  are real constants. The even kernel in the function is cosine modulated and odd kernel is sine modulated. The filter will have a real and imaginary components.

The inputs, state and output of an image is  $u_{ij}$ ,  $x_{ij}$  and  $y_{ij}$  respectively. In our architecture pipelined datapaths are used. If the input image pixel values and the initial states satisfy the equation

$$|u_{ij}| \leq 1, |x_{ij}^R(0)| \leq 1, |x_{ij}^I(0)| \leq 1 \text{ for } i=1 \dots M, j=1 \dots N.$$

Then the real and imaginary parts of the output image pixel values are kept in the same range.

From [6] and [7] we get the values of  $\alpha_x$ ,  $\alpha_y$ ,  $\beta_x$  and  $\beta_y$ .

The values of  $x^R$  and  $x^I$  are in between -1 and +1. The values of

$$\alpha_x = \frac{\cos \omega_x \lambda}{4 + \lambda^2}, \beta_x = \frac{\cos \omega_y \lambda}{4 + \lambda^2}, \alpha_y = \frac{\sin \omega_x \lambda}{4 + \lambda^2}, \beta_y = \frac{\sin \omega_y \lambda}{4 + \lambda^2}$$

$$b = \frac{\lambda^2}{4 + \lambda^2}$$

$\alpha_x$ ,  $\alpha_y$ ,  $\beta_x$ ,  $\beta_y$  and  $b$  are in between -0.25 and +0.25. these filter coefficients are changed according to the values of  $\omega_x$ ,  $\omega_y$  and  $\lambda$ .

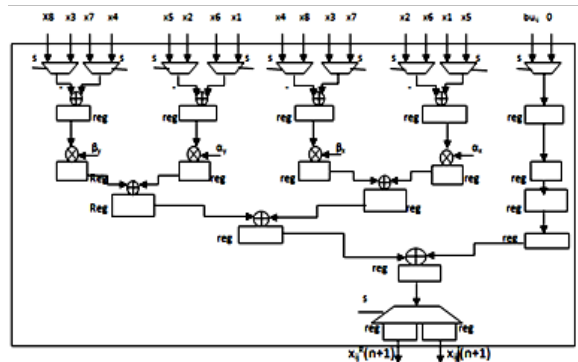
### Proposed Method

The fig1 shows the logical block diagram of a GTF(Gabor Type Filter) unit. Inputs for the given system are  $x_1, x_2, \dots, x_8$ . Another input is  $b$ , which is used for the determination of real part only.

For the multipliers the inputs are  $\beta_x$ ,  $\beta_y$ ,  $\alpha_x$ ,  $\alpha_y$ . From the reference papers we get  $\omega_x = 30$  degree,  $\omega_y = 60$  degree,  $\lambda = 0.5$ . By using these values we get  $\beta_x = 0.1176$ ,  $\beta_y = 0.2037$ ,  $\alpha_x = 0.2037$ ,  $\alpha_y = 0.1176$  and  $b = 0.05882$ . In this architecture we are using floating point addition and multiplication. Floating point computation is more precise and accurate than any other arithmetic computation.  $X$  is the state and it ranges from  $-1 < x < 1$ .

### Floating point conversion

Consider a 16 bit number. The 16 bit number is divided into 1 sign bit, 6 bit exponent part and 9 bit mantissa part.



1bit                  6bit                  9 bit

Fig1. Logical block diagram of gabor filter

1 100000 100000000

First value  $x_1 = -1$ , the sign bit is 1, because of negative value. The exponent part of -1 is obtained by the below procedure

$$-1 \rightarrow 1.00000$$

One right shift

$$1.0 \rightarrow 0.100000$$

Base value is  $2^{6-1} - 1 = 2^5 - 1 = 32 - 1 = 31$ . The exponent part is Number of shifting + Base value. Here the number of shift is 1. So the exponent is  $1 + 31 = 32 \rightarrow 100000$  6 bit exponent.

Next is the mantissa part which is the 9 bit value 100000000. Then the 16 bit floating point value for -1 is

Next value  $x_2 = 0.5$ . The binary value of 0.5 is 0.1. the sign bit is 0, because the number is positive. The exponent part is number of shift + base value. Here there is no shifting, so the exponent is the base value itself, ie  $2^{6-1} - 1 = 31 \rightarrow 11111$ , for 6 bit exponent 011111. The mantissa part is 100000000. Then the 16 bit floating point value of 0.5 is

0 011111 100000000

Next value is 0.25. Its binary value is 0.01. So one left shift is needed. Here the sign bit is 0, since 0.25 is a positive number. Because of left shifting the exponent value is (base value - 1), ie  $31 - 1 = 30$ , binary value is 11110. The exponent 6 bit value is 011110. The mantissa part is 100000000

0 011110 10000000

Next value is  $x_4 = 1$ . Here the sign bit is 1, because 1 is a positive value. The exponent and mantissa is same as that of  $x_1$ .

0 100000 100000000

Next we have to convert the values of  $\beta_x = 0.1176$ ,  $\beta_y = 0.2037$ ,  $\alpha_x = 0.2037$ ,  $\alpha_y = 0.1176$  and  $b = 0.05882$  to floating point numbers. First is the value of  $b$ , the binary value of  $b = 0.0588235 \rightarrow 0.0000111100001111$ .

For converting  $b$  to floating point value 4 left shifting is needed. The sign bit is 0. The exponent value is (base value - 4) =  $31 - 4 = 27$ . The binary value of 27 is 11011.the

mantissa part is 111100001.

0 011011 111100001

From  $x_1$  to  $x_8$

State	Sign bit	Exponent	Mantissa
X1= -1	1	100000	100000000
X2 = 0.5	0	011111	100000000
X3= 0.25	0	011110	100000000
X4= 1	0	100000	100000000
X5= -1	1	100000	100000000
X6= 0.5	0	011111	100000000
X7= 0.25	0	011110	100000000
X8= 1	0	100000	100000000

Table1

Next value is  $\alpha_x = 0.2037$ . the binary value of 0.2037 is 0.001101010110100111. For converting this value to floating point, 2 left shifting is needed. Sign bit is 0, the exponent value is (base value - 2) =  $31 - 2 = 29 \rightarrow 11101$ . The mantissa part is 110101011

0 011101 110101011

Since  $\beta_y = \alpha_x = 0.2037$ , the floating point value of  $\beta_y$  is same as  $\alpha_x$ , 0 011101 110101011.

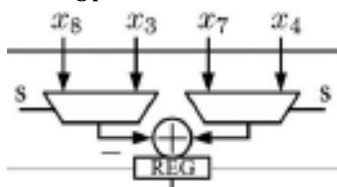
Next is the value of  $\beta_x = 0.1176$ . the binary value is 0.0001111000011110. For converting this to floating point value 3 left shifting is needed. Sign bit is 0, the exponent value is (base value - 3) =  $31 - 3 = 28 \rightarrow 11100$ . The mantissa part is 111100001

0 011100 111100001

The value of  $\beta_x = \alpha_y = 0.1176$ , so the floating point value of  $\alpha_y$  is same as that of  $\beta_x$ .

1 100000 011000000

**Floating point addition**



When the select line  $s = 0$ ,  $x_8$  and  $x_7$  is selected by the multiplexer. First  $x_8$  is multiplied by -1. So the value of  $x_8$  becomes 1 100000 100000000. The value of  $x_7$  is 0 011110 100000000.

$x_8$  1 100000 100000000

$x_7$  0 011110 100000000

$x_8 \rightarrow Y_1, x_7 \rightarrow Y_2$

$Y_3 = Y_1 + Y_2$

For addition of two floating point numbers make the exponent part same. The exponent part of  $x_8$ ,  $exp_1$  is 32 and the exponent part of  $x_7$ ,  $exp_2$  is 30. the difference of  $exp_1$  and  $exp_2$  ( $exp_1 - exp_2$ ) is  $32 - 30 = 2$ . Add 2 to  $exp_2$ , so that  $exp_1 = exp_2$ . since 2 is added to exponent part of  $x_7$ , the mantissa part of  $x_7$  is right shifted by 2 units, 100000000 becomes 001000000. Since the sign of two numbers are different, we have to subtract mantissa2 from mantissa1 (mantissa1 - mantissa2). Sign of  $Y_1 + Y_2$  is the sign of highest exponent, 1

Mantissa1 - mantissa2  $\rightarrow$  100000000 - 001000000 = 011000000

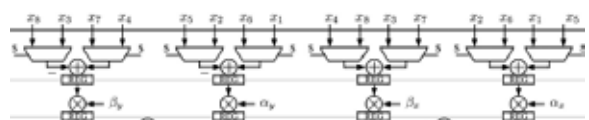
$Y_3 = Y_1 + Y_2$  is

Like this way when  $s = 0$ ,  $Y_1$   $x_3$ ,  $Y_2$   $x_4$ , then  $Y_3$  is  $Y_1 + Y_2$  ie  $x_3 + x_4$ .

1 100010 010100000

Addition operations for each multiplexers is carried out in this way and the values are stored in the registers. Then we have to do the multiplication operation of floating point numbers.

**Floating point multiplication**



Next we have to do the floating point multiplication. Consider the multiplier in the architecture. The inputs to the multiplier is  $Y_3 \rightarrow 1 100000 100000000$  and

$\beta_y \rightarrow 0 011101 110101011$ . The multiplier output is  $Y_4 = Y_3 * \beta_y$ .

The sign bit of the product is obtained by XOR operation. The sign bit of  $Y_4$  is, sign bit of  $Y_3$  (XOR) sign bit of  $\beta_y$ , ie  $1 \wedge 0 = 1$ . The sign bit of  $Y_4$  is 1.

Next is the exponent part.

Exp1- base value1 =  $32 - 31 = 1$

Exp 2- base value2 =  $32 - 29 = 2$

Therefore the final exponent is  $1 + 2 + 31 = 34$ , 100010

The mantissa part is obtained by normal multiplication of two binary numbers. The output of the multiplier,  $Y_4$  is

**Results**

The proposed structure for gabor filter is designed using floating point computation. The inputs given to the filter is  $x_1$  to  $x_8$ . The given input ranges from -1 to +1. These values are converted to floating point values using the verilog program. Then the addition operator adds the values and multiplier gives the product of the values given to it. Finally we will get the real and imaginary part of the image given to the gabor filter. The simulation is done using Xilinx for getting real and imaginary part. Finally the FPGA implementation is obtained on SPARTAN 3E.

**Conclusion and Future work**

In this brief, the GTF structure is proposed using floating point method. By introducing floating point computation we will get an accurate output for the given input. We can use real number

values instead of integers used in fixed point arithmetic. These floating point computations are done using the verilog codes. Finally the implementation of the GTF on FPGA is obtained.

Our future work includes finding the design details of the gabor filter using floating point computation and to compare these values with the fixed point operation.

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